

Supplementary Information: *Monolithic Ge/SiGe quantum well photodetectors integrated with SiN waveguides on 200 mm silicon wafers*

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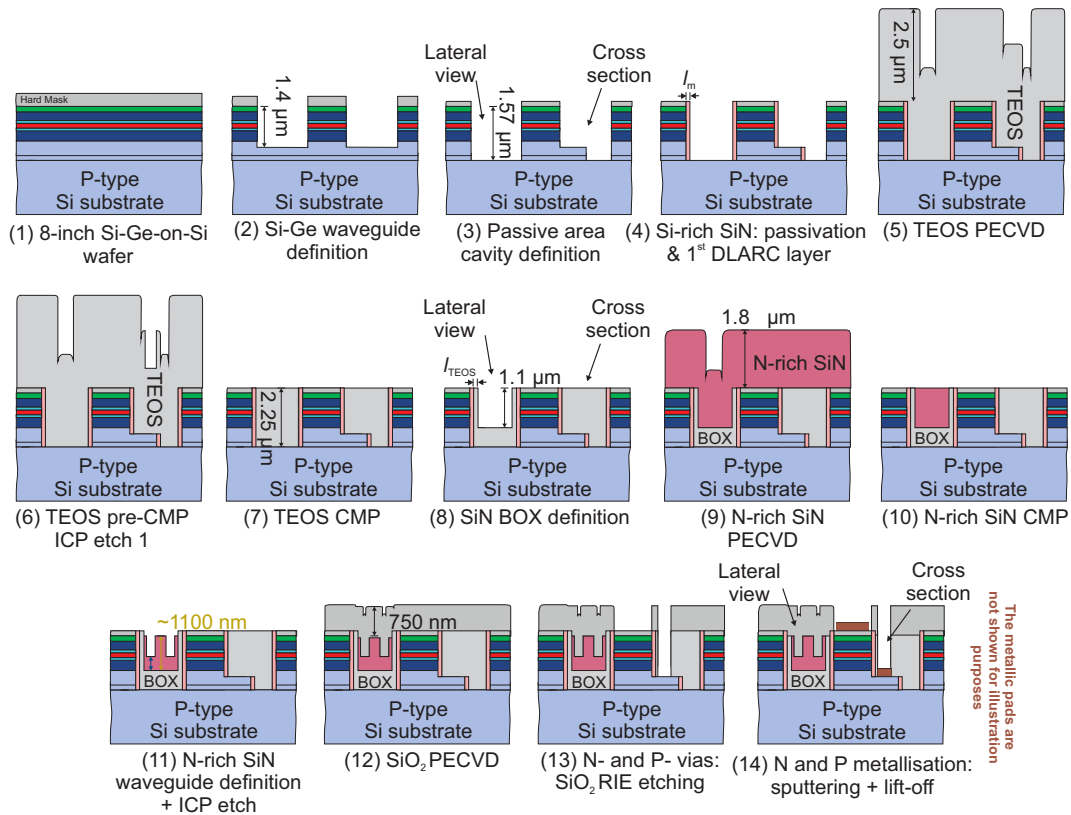
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I Fabrication Process Steps

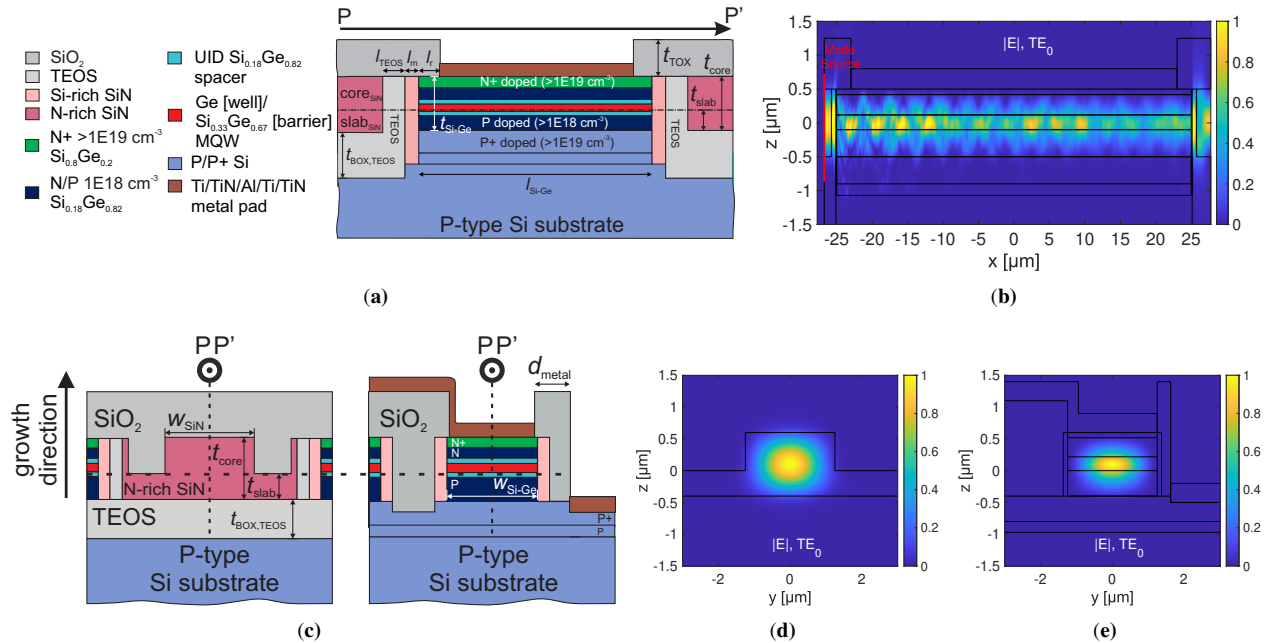
The fabrication process mitigates the drawbacks typically associated with the integration of active and passive waveguides, by embedding the amorphous passive waveguide layers within the Si-Ge cavities, instead of the conventional method of growing Si-Ge layers within a SOI cavity. This method effectively prevents the formation of defective Si-Ge layers at the boundary, leading to increased optical losses, reflections as well as detection uncertainty due to non uniformity in the MQW stack. Consequently, this approach eliminates the need for the processing of defective material from the interfaces. Instead, chemical-mechanical polishing (CMP) steps are implemented to enable the formation of amorphous waveguiding layers. This substitution ensures the precise control over the structure, position, thickness, and length of the waveguides and the layers at the interfaces, enhancing the overall reliability and performance of the active waveguide system. Supplementary Fig. 1 illustrates the fabrication steps for the photodetector integration.



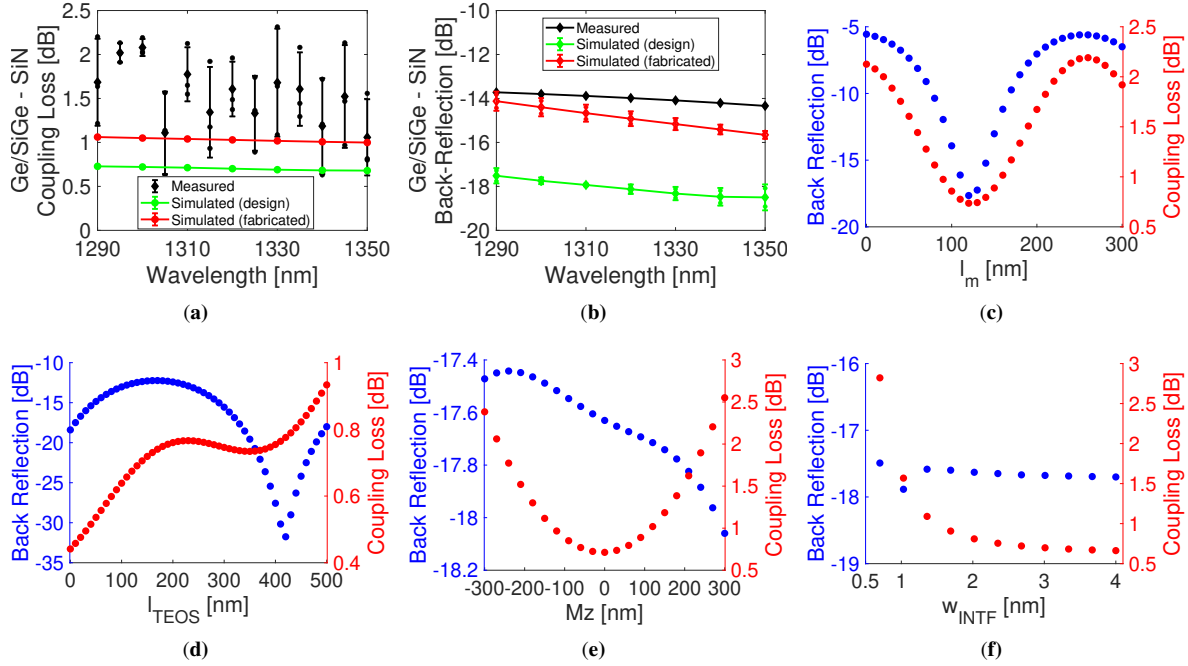
Supplementary Fig. 1. Fabrication process flow. The integration steps of a Ge/SiGe photodetector with a N-rich SiN waveguide on a Si substrate is shown.

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for fabricating the Si-Ge layer. The mean and standard deviation of the measured coupling losses were compared against simulations for both the ideal interface design and the actual fabricated junction. As shown in Fig. 3(a), the experimentally obtained range of coupling losses overlaps with the simulated response of the fabricated devices (red curve). The minimum measured loss occurred at 1305 nm, with a value of 1.10 ± 0.47 dB.



Supplementary Fig. 3. Coupling loss and back-reflection at the Ge/SiGe - SiN interface. (a) Measured coupling loss values compared with the simulated response for the optimally designed and the fabricated cases ($n=3$ independent points). (b) The back-reflection as measured accompanied by the corresponding modelled values for the optimally designed and the fabricated cases ($n=2$ independent points, S_{11} and S_{22}). The coupling performance is investigated under the variation of (c) the Si-rich SiN length, (d) the TEOS anti-reflective coating length, (e) the z-axis misalignment and (f) the waveguide width, with all the other parameters constant.

To quantify the reflection at each interface, a method based on cut-back structures was implemented. First, the propagation loss of the Si-Ge waveguides was determined using a dedicated non-metallised cut-back device. In parallel, cascaded junctions between N-rich SiN and non-metallised Si-Ge sections were fabricated in order to isolate and analyse the reflective behaviour of a single interface. The reflectivity of one interface was then extracted by fitting the measured optical spectra to numerical predictions obtained from a scattering-matrix model, following the approach in Supplementary Ref. 2. A comparison between the experimentally measured reflection spectrum and the simulated response, for both the designed and realised structures, is presented in Supplementary Fig. 3(b). The characterised reflection is close to the simulated response for the fabricated devices across the entire wavelength range.

A parametric study was conducted to evaluate the sensitivity of the coupling mechanism through individual sweeps of key geometric parameters. Specifically, the coupling loss and back-reflection of a single waveguide interface were simulated under variations in the interface geometry, as illustrated in Fig. 3(c)–3(f). The parameters investigated include the lengths of the Si-rich SiN (l_m) and TEOS (l_{TEOS}) anti-reflective layers, the vertical misalignment between the waveguides (M_z), and the interface width (w_{INTF}).

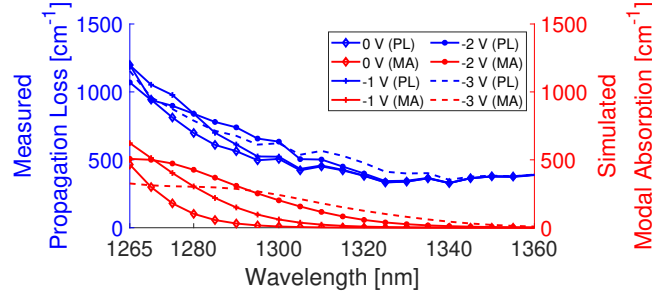
A deviation in the Si-rich SiN layer length from its optimal value compromises both the coupling loss and back-reflection, as the anti-reflective condition is no longer satisfied. Variations in the TEOS layer length have a similar impact on back-reflection; however, increasing l_{TEOS} expands the propagating mode, further diminishing the periodic coupling efficiency. The coupling performance is also highly sensitive to vertical alignment, exhibiting a minimum loss at zero displacement and a symmetric increase with increasing M_z due to misalignment of the waveguide centres.

Finally, narrower interface regions lead to higher coupling losses owing to the intensified mismatch in confinement factors between the high-index (Si-Ge) and mid-index (SiN) waveguides. This effect is progressively mitigated as the interface width increases and the optical fields become more spatially localised within the respective waveguides, approaching an asymptotic

71 coupling limit.

72 III Losses

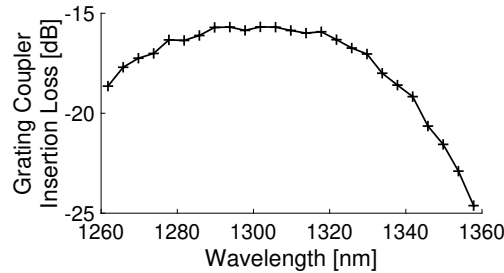
73 The total propagation losses of the 50 μm long Si-Ge photodetector were extracted by measuring its insertion loss followed
 74 by a normalisation with a silicon nitride reference waveguide. These include the QCSE absorption, the Ge and SiGe indirect
 75 absorption, the doped Si absorption, the Si-rich SiN absorption, the losses due to the metal pads and the scattering loss of the
 76 waveguide. The modal absorption by the active layers was simulated by using the package Harold EAM from the commercial
 77 software Photon Design. Both measured propagation losses and simulated modal absorption are plotted in Supplementary Fig.
 78 4 for the wavelengths 1265-1360 nm and for biases spanning from -3 to 0 V.



Supplementary Fig. 4. Propagation loss (PL) and modal absorption (MA).

79 IV Grating coupler

80 The photodetectors received the optical power through the SiN waveguides, which were connected with single-mode fibres
 81 via grating couplers with straight gratings. The grating structures were designed for operation at 1310 nm and comprised 30
 82 teeth, each 10 μm wide, with a period of 1016 nm and a 50% duty cycle. The insertion loss of one grating coupler is plotted in
 83 Supplementary Fig. 5.

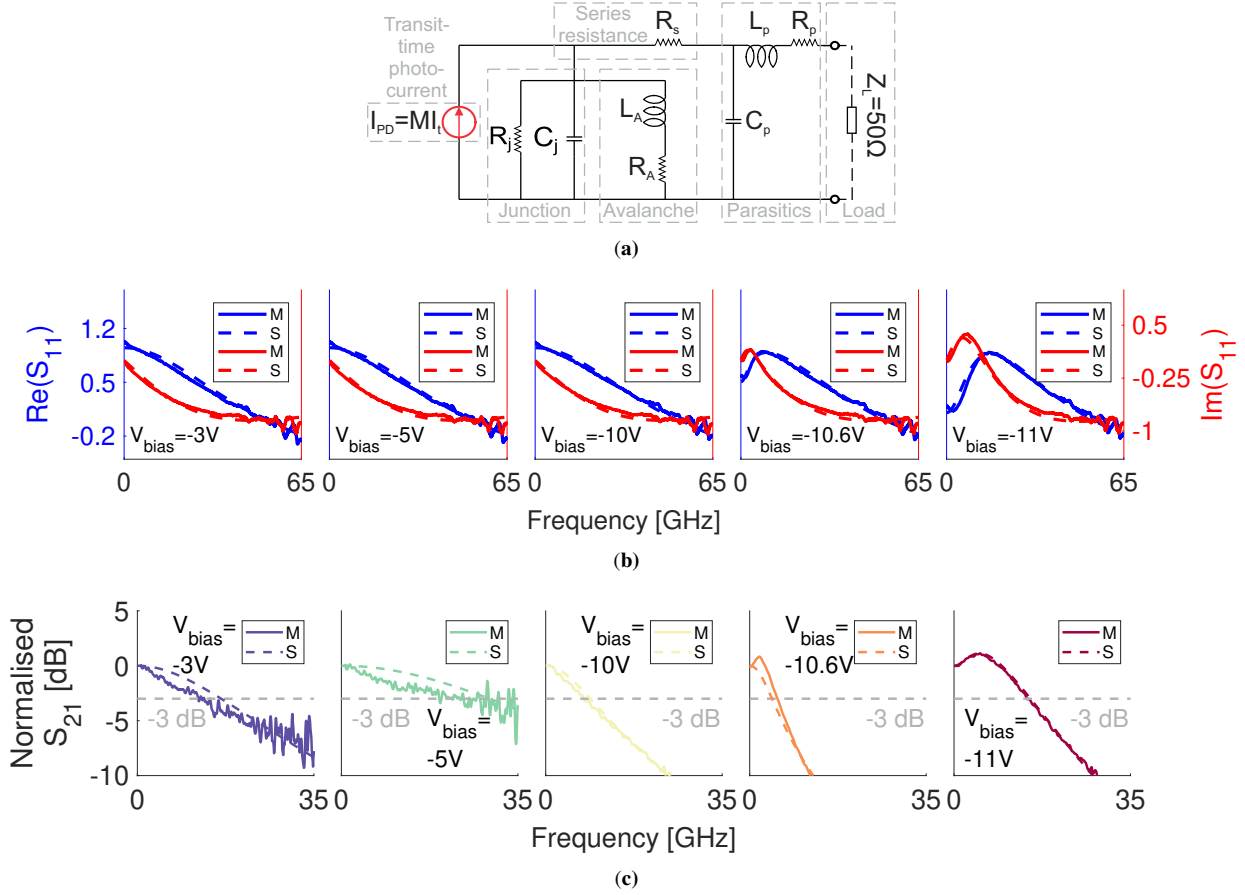


Supplementary Fig. 5. Grating coupler response.

84 V Radio Frequency Equivalent Model

85 The photodetector behaviour under both low-bias and avalanche conditions is described using an equivalent circuit model
 86 as shown in Supplementary Fig. 6(a). The device is represented by a current source with a single-pole frequency response
 87 (RC equivalent) that includes the total carrier delay and multiplication factor M . The total delay $\tau_{\text{total}} = \tau_{\text{tr}} + \tau_{\text{m}}$ combines the
 88 carrier transit time (τ_{tr}) and the avalanche build-up time (τ_{m}). At low bias, only τ_{tr} is relevant and $M = 1$. The transfer function
 89 of the current source is expressed as $H(\omega) = \frac{1}{1 + j\omega M \tau_{\text{total}}}$. The multiple quantum well region is modelled by its junction
 90 resistance (R_j) and capacitance (C_j). In avalanche mode, an inductance (L_A) and resistance (R_A) are added to account for the
 91 phase delay between the applied high-frequency voltage and the photocurrent generated (space charge effect). The circuit also
 92 includes a series resistance (R_s) and parasitic components (R_p, L_p, C_p), terminated with a 50 Ω load.

93 Model parameters were obtained by fitting the simulated S_{11} response to one-port measurements, enabling accurate
 94 prediction of the S_{21} response at 1287 nm, as shown in Supplementary Fig. 6(b) and 6(c), respectively. Measurements at biases
 95 from -3 V to -11 V are in close agreement. A resonance between L_A and C_j above -10.6 V produces a gain peak and a



Supplementary Fig. 6. High frequency parameters. (a) The radio frequency (RF) equivalent circuit of the Si-Ge photodetector. (b) The measured and fitted real and imaginary parts of the S_{11} parameter. (c) The measured S_{21} response of the photodetector overlaid with the simulated curves using the fitted circuit parameters. M: measured, S: simulated.

bandwidth enhancement. The bandwidth decreases near -10 V, reaches a minimum around -10.6 V, and increases again due to the resonance-assisted gain.

Supplementary Table 1. Fitted parameters of the radio-frequency equivalent circuit.

V_{bias} [V]	R_j [Ω]	C_j [fF]	L_A [nH]	R_A [Ω]	R_s [Ω]	L_p [pH]	R_p [Ω]	C_p [fF]	M	τ_{total} [ps]
-3	3100	39	∞	∞	9	4	3.3	21	1	8.5
-5	2800	39	∞	∞	9	4	3.3	21	1	3.69
-10	2680	39	∞	∞	9	4	3.3	21	5.3	3.24
-10.6	1080	39	7.8	124	9	4	3.3	21	10.5	5
-11	3000	39	2.1	50	9	4	3.3	21	8.5	0.73

The fitted circuit parameters are noted in Supplementary Table 1. For a low bias < -10 V the part represented by the inductor L_A and the R_A is considered as an open circuit. When the voltage is high enough for the avalanche effect to occur, the L_A and R_A get decreasing finite values. The junction resistance R_j follows the photocurrent and voltage gradients through the expression $R_j = \left(\frac{dV}{dI}\right)^{-1}$ and can be estimated by the IV curve and further fitting. Negligible changes are assumed for the capacitance of the fully depleted junction for voltages higher than -3 V, while the series resistance and the circuit parasitics are considered constant against the bias variation. The multiplication factor M is calculated by normalising the generated photocurrent with the reference value at -5 V. Following the above calculations, the total time τ_{total} is calculated based on the

105 fitting of the S_{21} parameter.

106 **Supplementary References**

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